

# International Journal of Sciences: Basic and Applied Research (IJSBAR)

International Journal of
Sciences:
Basic and Applied
Research
ISSN 2307-4531
(Print & Online)
Published by:

**ISSN 2307-4531** (Print & Online)

http://gssrr.org/index.php?journal=JournalOfBasicAndApplied

A Novel Higher-Order Sliding Mode Control Scheme for Uncertain Nonlinear Systems: Short-period Missile Control Application

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#### **Abstract**

The paper proposes a novel higher-order sliding modes (HOSM) control scheme for a class of uncertain nonlinear systems. The HOSM-based control scheme is developed based on the Filippov's differential inclusion and local properties of affine nonlinear systems with control constraints. The resulting control provides exponential stability and ensures robustness against modeling errors and parameter uncertainties. The proposed HOSM-based control scheme is used to design a short-period pitch-axis flight control system of a short-range tactical missile where performance and robustness are demonstrated via computer simulations.

*Keywords:* Flight control system design; Higher-order sliding modes; Missile dynamic and control; Nonlinear robust control

#### 1. Introduction

For high tracking performance, the design of a missile's flight control system (FCS) is driven by the characteristics of the guidance commands which are determined by the overall system and homing loop requirements. In many interception and navigation applications, the selection of the performance index changes with the flight phase of the

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missile which requires the change of the autopilot type within the FCS. Because of the short amount of time involved in the endgame interception phase, the basic requirements for a tactical missile's FCS to achieve a successful hit-to-kill engagement are fast response, minimum error, and robustness against disturbances.

In order to alleviate the problems associated to the classical linear autopilot and to respond to the requirements of successful endgame engagement against clever targets, various studies have focused on the problem of nonlinear autopilot design for tactical missiles during the last few years. Researchers have sought to augment the classical autopilots with modern robust controllers yielding many control FCSs schemes [1-10]. Although many of these FCSs ensure good tracking, their application is restrained due to certain conceptual and implementation shortcomings such as linearization, high computational cost, and vulnerability to disturbances.

Since their development, Higher-Order Sliding Modes (HOSM) have been receiving more attention and study from aeronautical and aerospace communities. Having proved their high accuracy and robustness, HOSM have been increasingly used to design high-performance FCSs for advanced missiles [11-22]. HOSMs controllers mitigate the problems associated with standard sliding mode controllers such as high-order dynamics, chattering effect, and control input smoothness.

In this paper a new nonlinear discontinuous HOSM control scheme is derived and used to design an advanced HOSM-based pitch-axis tactical missile FCS. Different from the recursive or nested algorithms presented in [23,24], the proposed approach uses directly the higher-order time derivatives of the sliding variable in one combination. This reduces the complexity of the controllers' architecture for systems with higher relative degrees. Two different pitch-axis autopilots are designed; pitch-rate and angle-of-attack (AOA) autopilots. The performance requirements of these autopilots are demonstrated via computer simulations.

The outline of this paper is as follows. The nonlinear short-mode missile's dynamics are presented in section 2. In section 3 the problem of designing discontinuous HOSM-based control is stated and the proposed HOSM scheme is derived. Computer simulations are conducted in section 4 to demonstrate the efficiency and advantages of the proposed HOSM-based missile's FCS. Section 5 concludes the paper.

#### 2. Missile Longitudinal Short-period Dynamics

The short period mode of the longitudinal motion can be described as follows

Missile short-mode dynamics

$$\begin{cases} \dot{\alpha} = K_{\alpha} M C_n(\alpha, M) cos(\alpha) + q + \frac{g}{v_s M} cos(\gamma) \\ \dot{q} = K_q M^2 C_m(\alpha, M) + K_q M^2 e_m q + d_m \delta \end{cases}$$
 (1)

where the aerodynamic coefficients  $C_n$  and  $C_m$  are estimated as follows

$$\begin{cases}
C_n = sgn(\alpha) \left[ a_n |\alpha|^3 + b_n \alpha^2 + c_n (2 - M/3) |\alpha| \right] \\
C_m = sgn(\alpha) \left[ a_m |\alpha|^3 + b_m \alpha^2 + c_m (-7 + 8M/3) |\alpha| \right]
\end{cases}$$
(2)

Missile velocity and flight patch angle

The missile velocity and trajectory are introduced by means of Mach number M and path angle  $\gamma$  as exogenous scheduling parameters generated by the following equations

$$\begin{cases} \dot{M} = -K_z M^2 \left[ C_{D_0} - C_z(\alpha, M) sin\alpha \right] - \frac{g}{v_s} sin(\gamma) \\ \dot{\gamma} = -K_Z M C_Z(\alpha, M) cos(\alpha) - \frac{g}{v_s M} cos(\gamma) \end{cases}$$
(3)

• Missile normal acceleration

$$\eta = K_z M^2 C_n(\alpha, M) \tag{4}$$

• Actuator dynamics:

$$\ddot{\delta} = -\omega_a^2 \delta - 2\xi_a \omega_a \dot{\delta} + \omega_a^2 \delta_c \tag{5}$$

The state-space representation of the model (3) is written as follows

$$\begin{cases}
\dot{\mathbf{x}} = \mathbf{A}(\mathbf{x})\mathbf{x} + \mathbf{b}\mathbf{u} \\
\mathbf{y} = \mathbf{c}^T \mathbf{x}_{\mathbf{d}}
\end{cases} \tag{6}$$

where

$$\begin{cases}
\mathbf{x} = [x_1 \quad x_2]^T = [\alpha \quad q]^T, \quad u = \delta \\
A(\mathbf{x}) = \begin{bmatrix} A_{11}(\mathbf{x}) & 1 \\ A_{21}(\mathbf{x}) & 0 \end{bmatrix}, \quad \mathbf{b} = \begin{bmatrix} 0 & d_m \end{bmatrix}^T \\
A_{11}(\mathbf{x}) = sgn(x_1)K_{\alpha}M[a_nx_1^2 + b_n|x_1| + c_n(2 - M/3)]cos(x_1) \\
A_{21}(\mathbf{x}) = sgn(x_1)K_{\alpha}M^2[a_mx_1^2 + b_m|x_1| + c_m(-7 + 8M/3)]
\end{cases} (7)$$

 $A_{11}(\mathbf{x})$  and  $A_{21}(\mathbf{x})$  are uncertain sufficiently smooth functions and  $\mathbf{x}_d = [\alpha_d(t) \quad q_d(t)]^T$  is the desired output vector. The variables  $-20^\circ \le \alpha \le 20^\circ$ ,  $1.5 \le M \le 3$ , q,  $\gamma$ , and  $\delta$  denotes Angle-Of-Attack (AOA), Mach number, pith rate, path angle, and elevator deflection, respectively. The parameters  $v_s = 1036.4 \, (ft/s)$ ,  $P_0 = 973.3 \, (lb/ft^2)$ ,  $I_y = 182.5 \, (slug.ft^2)$ ,  $S = 0.44 \, (ft^2)$ ,  $d = 0.75 \, (ft)$ ,  $m = 13.98 \, (slug)$ ,  $\omega_a = 150 \, (rd/s)$ , and  $\xi_a = 0.7$  denotes sound speed, static pressure, moment of inertia with respect to pitch axis, reference area, reference diameter, airframe mass, actuator natural frequency, and actuator damping ratio, respectively. Numerical values of aerodynamic coefficients of a pitch-axis missile model at an altitude of 20,000 ft are listed in Table 1 [7].

Table 1. Aerodynamic coefficients

| Coefficient | Value          |
|-------------|----------------|
| $K_z$       | $0.7P_0S/mv_s$ |
| $K_m$       | $0.7P_0Sd/I_y$ |
| $a_n$       | 19.373         |
| $b_n$       | -31.023        |
| $c_n$       | -9.717         |
| $C_{D0}$    | -0.30          |
| $a_m$       | 40.434         |
| $b_m$       | -64.015        |
| $c_m$       | 2.922          |
| $d_m$       | -11.803        |
| $e_m$       | -1.719         |

The control objectives are to design an inner-loop feedback controller with  $c = [0 \ 1]^T$  to track a pitch rate command  $q^c(t)$  and an outer-loop feedback controller with  $c = [1 \ 0]^T$  to track an AOA command  $\alpha^c(t)$ .

#### 3. Discontinuous HOSM-based Control Design

#### 3.1. Problem statement

Consider a certain class of dynamical systems characterized by smooth nonlinear dynamics and represented by the following closed-loop state-space feedback

$$\begin{cases}
\dot{\mathbf{x}} = \mathbf{A}(\mathbf{x})\mathbf{x} + \mathbf{b}(\mathbf{x})\mathbf{u} \\
\mathbf{y} = \sigma(\mathbf{x})
\end{cases} \tag{9}$$

where  $x \in \mathcal{R}^n$ ,  $\sigma \in \mathcal{R}$ , and  $u \in \mathcal{R}$ . The functions  $A_{ij}$  and  $b_i$  are uncertain continuous sufficiently differentiable functions. Using discontinuous higher-order sliding mode (HQSM) control approach, the control problem considered in this study aims to design high-performance dynamic controllers for real-time tracking missions in spite of internal and external disturbances. The discontinuous HQSM-based control approach is formulated under the following assumptions [23,24].

Assumption 1: The matrix A(x) and the vector b(x) are partitioned into nominal and uncertain parts as follows

$$\begin{cases} A(x) = A_{nom}(x) + \Delta A(x) \\ b(x) = b_{nom}(x) + \Delta b(x) \end{cases}$$
(10)

with

$$\begin{cases} \|\Delta A(x)\| \le A_{max} \triangleq \max_{x \in \mathbb{X}} |\Delta A(x)| \in \mathbb{R}^+ \\ \|\Delta b(x)\| \le b_{max} \triangleq \max_{x \in \mathbb{X}} |\Delta b(x)| \in \mathbb{R}^+ \end{cases}$$
(11)

where X is the oerationg space of the system (9).

Assumption 2: The control input u is a Lebesgue-measurable signal (affine scalar function) with

$$|u(t)| \le u_{max} \in \mathbb{R}^+ \tag{12}$$

Assumption 3: The output constraint  $\sigma(x)$  is of r-order differentiability class where r defines the relative degree of the output constraint  $\sigma(x)$  with respect to the control input u [25].

Assumption 4: the output function  $\sigma(x)$  and its successive time derivatives up to r-1 form a non-empty integral set

$$\sigma^{r} = \left\{ \mathbf{x} \in \mathbb{X} | \ \sigma(\mathbf{x}) = \dot{\sigma}(\mathbf{x}) = \dots = \sigma^{(r-1)}(\mathbf{x}) = 0 \right\} \neq \emptyset - integral \ set$$
 (13)

Assumption 5: if the assumption (3) holds and r is constant and known, the rth- order time derivative of the output constraint  $\sigma(x)$  satisfies the following equality [25]

$$\sigma^{(r)} = p(\mathbf{x}) + q(\mathbf{x})u \tag{14}$$

where  $p(\mathbf{x}) = L_f^r \sigma(\mathbf{x})$  and  $q(\mathbf{x}) = L_g L_f^{r-1} \sigma(\mathbf{x}_o) \neq 0$ . Due to the assumption 1, the functions  $p(\mathbf{x})$  and  $q(\mathbf{x})$  are bounded continuous functions on an open set  $\mathcal{X}(\mathbf{x}_o) \subset \mathbb{X}$  around a given initial condition  $\mathbf{x}_o$  with

$$0 < \rho_{-} \le q(x) \le \rho_{+} \qquad |p(x)| \le \vartheta \tag{15}$$

 $\rho_-$ ,  $\rho_+$  and  $\vartheta$  are some positive constants.

Assumption 6: According to the geometric-differential theoretical framework [25], the system (9) is feedback equivalent under a local diffeomorphic coordinate transformation  $z_k = \sigma^{(k)}$  (k = 0, ..., r). A controllable Brunovsky canonical form can be constructed as follows

$$\dot{\mathbf{z}} = \mathbf{\Lambda}\mathbf{z} + \mathbf{\Upsilon}u \tag{16}$$

where

$$\begin{cases} \mathbf{\Lambda} = \begin{bmatrix} L_f \sigma(\mathbf{x}) & L_f^2 \sigma(\mathbf{x}) & \dots & L_f^{r-1} \sigma(\mathbf{x}) & L_f^r \sigma(\mathbf{x}) \end{bmatrix}^{\mathrm{T}} \\ \mathbf{\Upsilon} = \begin{bmatrix} 0 & 0 & \dots & 0 & L_g L_f^{r-1} \sigma(\mathbf{x}_o) \end{bmatrix}^{\mathrm{T}} \end{cases}$$
(17)

#### 3.2. Design of discontinues r-HQSM controllers

Theorem 1 [24]: Let assumptions 1-6 hold, a bounded r-sliding mode feedback controller such as

$$u_r = -G\Psi_{r-1,r}(\sigma, \dot{\sigma} \dots, \sigma^{(r-1)})$$
(18)

can be constructed in the constraint-space (16) to drive the constraint output  $\sigma(x)$  and its successive (r-1) time derivatives towards their zero level in finite time and in spite of uncertainties and disturbances.

**Theorem 2:** Let  $K_1, ..., K_r > 0$ , the following higher-order sliding mode-based controller

$$\begin{cases} u_r = -G\Psi_r(\sigma, \dot{\sigma} \dots, \sigma^{(r)}) \\ \Psi_r = \sigma^{(r)} + K_1 |\sigma^{(r-1)} + \sum_{i=r-2}^0 K_i \sigma^{(i)}| * sign(\sigma^{(r-1)} + \sum_{i=r-2}^0 K_i \sigma^{(i)}) \end{cases}$$
(19)

stabilizes the origin of the system (16) and ensure the establishment of the r-sliding mode  $\sigma \equiv 0$  in finite time provided that the gain G and the design parameters  $K_i$  are selected such that

$$G > \vartheta/\rho_{-}(|\sigma^{(r)}| + K_1|\sigma^{(r-1)} + \sum_{i=r-2}^{0} K_i \sigma^{(i)}|)$$
 (20)

The controller (19) can be implemented using one of the following saturation functions

$$\Psi_r = \begin{cases} \Psi_r / \Delta & \text{if } |\Psi_r| \le \Delta \\ \Psi_r = \text{sign}(\Psi_r) & \text{if } \Psi_r > \Delta \end{cases}$$
 (21)

or

$$\Psi_r = \Delta - \exp(-\mu \Psi_r) \tag{22}$$

where  $\Delta$  and  $\mu$  are convergence parameters.

**Proof of theorem 2:** Define a sliding function  $s = \sigma^{(r-1)}$  and consider the following Lyapunov candidate function

$$V = \frac{1}{2}S^T S \tag{23}$$

Using the equations (14) and (19), the first-order time-derivative of the function V is given as follows

$$\dot{V} = s^{T} \dot{s} = s^{T} \sigma^{(r)} 
= \sigma^{T} c^{T} [p(x) + q(x)u] 
= \sigma^{T} c^{T} \left[ p(x) - Gq(x) \left( \sigma^{(r)} + K_{1} \middle| \sigma^{(r-1)} + \sum_{i=r-2}^{0} K_{i} \sigma^{(i)} \middle| * sign(\sigma^{(r-1)} + \sum_{i=r-2}^{0} K_{i} \sigma^{(i)}) \right) \right] 
\leq |\sigma^{T}| c^{T} \left[ \vartheta - G\rho_{-} \left( |\sigma^{(r)}| + K_{1} \middle| \sigma^{(r-1)} + \sum_{i=r-2}^{0} K_{i} \sigma^{(i)} \middle| \right) \right] 
\leq -|\sigma^{T}| c^{T} \vartheta \left[ \frac{G\rho_{-} \left( |\sigma^{(r)}| + K_{1} \middle| \sigma^{(r-1)} + \sum_{i=r-2}^{0} K_{i} \sigma^{(i)} \middle| \right)}{\vartheta} - 1 \right] 
\leq -\eta |\sigma^{T}|$$
(24)

# 4. HOSM-based Missile's Longitudinal Autopilots Design and Validation

In this section, the proposed HOSM-based control scheme is validated through different nonlinear numerical simulations. For all the scenarios, the sliding variable  $\sigma(x)$  is selected as

$$\sigma(\mathbf{x}) = \mathbf{c}^T \mathbf{e} = \mathbf{c}^T (\mathbf{x} - \mathbf{x}_d) \tag{25}$$

with  $c \in \mathbb{R}^n$ . The identity (14) is fulfilled as follows

$$\sigma^{(r)} = \mathbf{c}^T \mathbf{e}^{(r)} = p(\mathbf{x}) + q(\mathbf{x})u \tag{26}$$

#### 4.1. Outer-loop AOA autopilot

The control task here is to enforce the missile's airframe to follow a desired AOA path using the nominal form of the dynamic model (6). According to the assumption 3, from the model (6) it is easy to find that the relative degree of the dynamics  $u \to \sigma(x) \equiv \delta \to \alpha$  is r = 2 which yields

$$\sigma^{(2)} = \begin{bmatrix} 1 & 0 \end{bmatrix} \begin{bmatrix} \alpha^{(2)} \\ q^{(2)} \end{bmatrix}$$

$$= \left( K_{\alpha} M \frac{\partial C_{n}}{\partial \alpha} \cos(\alpha) - K_{\alpha} M C_{n} \sin(\alpha) \right) \left( K_{\alpha} M C_{n} \cos(\alpha) + q + \frac{g}{v_{s} M} \cos(\gamma) \right)$$

$$+ K_{q} M^{2} C_{m} + K_{q} M^{2} e_{m} q - \alpha_{d}^{(2)}(t) + d_{m} \delta$$

$$= p(\mathbf{x}) + q(\mathbf{x}) \delta$$
(27)

The controller (19) is given as follows

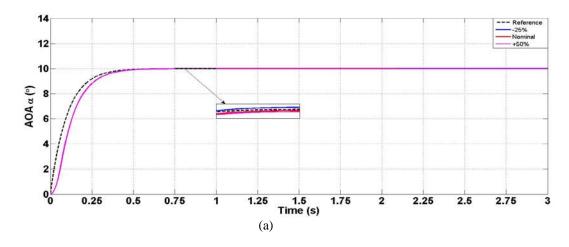
$$\begin{cases} u_2 = -G\Psi_2(\sigma, \dot{\sigma}, \sigma^{(2)}) \\ \Psi_2 = \sigma^{(2)} + K_1 |\dot{\sigma} + K_0 \sigma| * sign(\dot{\sigma} + K_0 \sigma) \end{cases}$$
(28)

#### Scenario 1: tracking of an asymptotic path

In this scenario, the performance and robustness of the proposed HOSM-based control scheme are evaluated through a smooth tracking of an asymptotic path. Nominal-, 25% underestimated-, and 50% overestimated-form of the model (6) were used in this simulation as shown in Figure 1.

#### Scenario 2: tracking of sinusoidal path

The control objective of this scenario is to track a sinusoidal reference trajectory at the extreme AOA conditions with  $\alpha_d(t) = 20 \sin(2t)$  (deg). Figure 2 shows the time-history of the AOA response and its corresponding control effort.



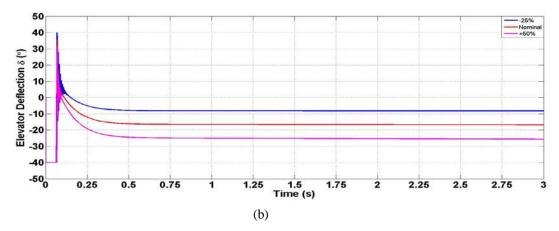


Figure 1. Time-histories of a tracking of asymptotic path scenario: (a) AOA response, (b) corresponding aerodynamic control.

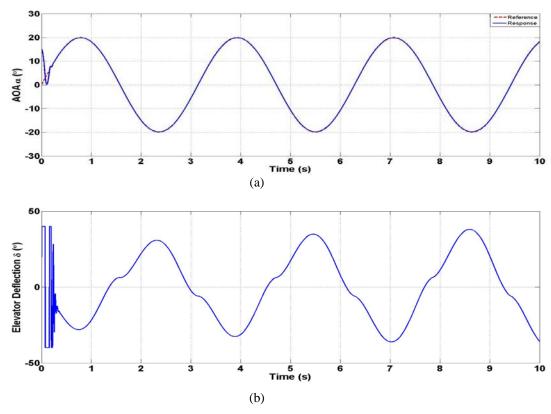


Figure 2. Time-histories of a tracking of a sinusoidal path scenario: (a) AOA response, (b) corresponding aerodynamic control.

# 4.2. Inner-loop pitch rate autopilot

It the case of pitch rate control, the missile's airframe is enforced to follows a desired pitch rate path as inner-loop control. According to the assumption 3, from the model (6) the relative degree of the dynamics  $u \to \sigma(x) \equiv \delta \to q$  is r = 1 which results in

$$\sigma^{(1)} = \begin{bmatrix} 0 & 1 \end{bmatrix} \begin{bmatrix} \alpha^{(1)} \\ q^{(1)} \end{bmatrix}$$

$$= K_q M^2 C_m + K_q M^2 e_m q + d_m \delta$$

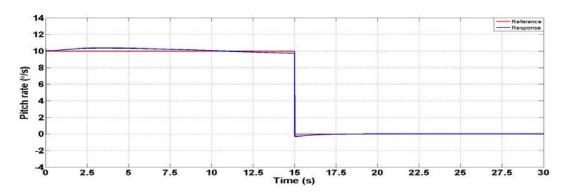
$$= p(x) + q(x) \delta$$
(29)

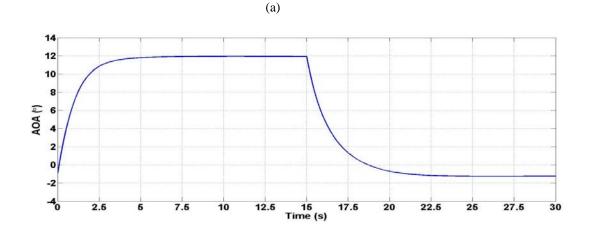
The controller (19) is given as follows

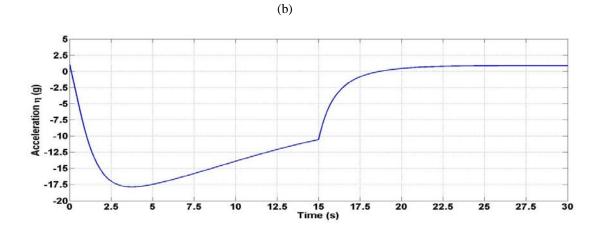
$$\begin{cases}
 u_1 = -G\Psi_2(\sigma, \dot{\sigma}) \\
 \Psi_1 = \sigma^{(1)} + K_1 |K_0 \sigma| * sign(K_0 \sigma)
\end{cases}$$
(30)

# Scenario 3: tracking of pitch rate command

Using the nominal form of the model (6), the missile's airframe is enforced to track a pitch rate pattern as shown in Figure 3.







(c)

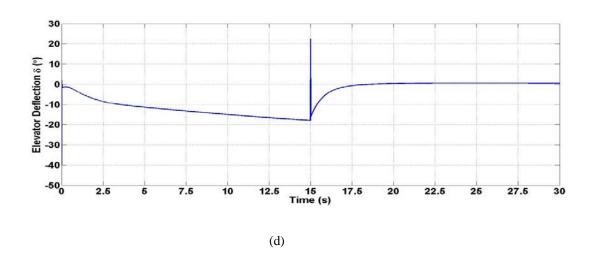


Figure 3. Time-histories of a tracking of a pitch rate pattern scenario: (a) pitch rate response, (b) AOA response, (c) Normal acceleration latex, (d) aerodynamic control.

#### 5. Conclusion

A robust higher-order sliding mode (HOSM) control scheme for nonlinear uncertain systems driven by an affine control input is proposed. The main contributions of this paper are the design of robust finite-time convergent HOSM-based controllers without resort to recursive procedures and chattering effect reduction even for first-order dynamics. The proposed HOSM-based control scheme was applied to designing pitch-axis autopilots for a short-range tactical missile during the endgame interception. High-tracking precision and robustness against heavy uncertainty conditions were achieved while the control input was maintained smooth.

Similar to many existing discontinuous sliding mode control algorithms, the proposed HOSM-based control scheme does not require an exact dynamic model of the controlled system or process. The only requirements for real-time implementation are the measurements of the sliding variable, the relative degree of the system output with respect to its input, and the bounds or restrictions (15). The unbounded uncertainties or disturbances, undefined relative degree, and high measurement noises remain unsolved problems and main challenges for the extension of the proposed research.

## Acknowledgments

This article was funded by the Deanship of Scientific Research (DSR), King Abdulaziz University, Jeddah. The author, therefore, acknowledges with thanks DSR technical and financial support.

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